

## Solutions of the First Two Tasks, CTH'08

• **Problem 1**

- a) Define the difference between a MIMO and a SISO control system with respect to the design of the feedback gains.

**Solution:** The feedback gain of MIMO system is a  $m \times n$  or  $m \times r$  matrix, depending on the types of the feedback control. The feedback gain of SISO system is a  $1 \times 1$  or  $m \times 1$  matrix.

The design of SISO feedback control can be formed by hand in the  $1 \times 1$  case, while the MIMO design is a full mathematical approach.

- b) State the state space description of the SISO system with the transfer function

$$F(s) = \frac{Y(s)}{U(s)} = k \frac{s^2 + 5}{(s^2 + 2s + 5)(s - 1)}$$

in state space notation.

**Solution:** First,

$$F(s) = \frac{k(s^2 + 5)}{(s^2 + 2s + 5)(s - 1)} = \frac{k(s^2 + 5)}{s^3 + s^2 + 3s - 5}.$$

Denote

$$\begin{aligned} a_0 &= 1, & b_0 &= 0, \\ a_1 &= 1, & b_1 &= k, \\ a_2 &= 3, & b_2 &= 0, \\ a_3 &= -5, & b_3 &= 5k, \end{aligned}$$

and define the state vector  $x(t) = [x_1 \ x_2 \ x_3]^T$  as

$$\begin{aligned} x_1 &= y - \beta_0 u, \\ x_2 &= \dot{x}_1 - \beta_1 u, \\ x_3 &= \dot{x}_2 - \beta_2 u, \end{aligned}$$

where

$$\begin{aligned} \beta_0 &= b_0 = 0, \\ \beta_1 &= b_1 - a_1 \beta_0 = k, \\ \beta_2 &= b_2 - a_1 \beta_1 - a_2 \beta_0 = -k, \\ \beta_3 &= b_3 - a_1 \beta_2 - a_2 \beta_1 - a_3 \beta_0 = 3k. \end{aligned}$$

The state space expression of the system equation should be

$$\begin{aligned} \dot{x}(t) &= Ax(t) + Bu(t), \\ y(t) &= Cx(t) + Du(t). \end{aligned}$$

where

$$A = \begin{bmatrix} 0 & 1 & 0 \\ 0 & 0 & 1 \\ -a_3 & -a_2 & -a_1 \end{bmatrix} = \begin{bmatrix} 0 & 1 & 0 \\ 0 & 0 & 1 \\ 5 & -1 & -1 \end{bmatrix}, \quad B = \begin{bmatrix} \beta_1 \\ \beta_2 \\ \beta_3 \end{bmatrix} = k \begin{bmatrix} 1 \\ -1 \\ 3 \end{bmatrix},$$

$C = [1 \ 0 \ 0]$ , and  $D = 0$ .

- c) State the conditions for the matrices  $P$ ,  $Q$  of the Lyapunov equation  $A^T P + P A = -Q$  to guarantee the asymptotic stability of the system  $\dot{x} = Ax(t)$ .

**Solution:**  $A^T P + P A = Q$  should be solvable and both  $P$  and  $Q$  should be symmetric positive definite matrices.

- d) For the system  $\dot{x} = Ax$  with the system matrix

$$A = \frac{1}{2} \begin{pmatrix} -1 & 1 & 0 \\ -1 & 1 & 4 \\ -1 & -1 & -2 \end{pmatrix}$$

and the matrices

$$P = \begin{pmatrix} 2 & 0 & 0 \\ 0 & 1 & 1 \\ 0 & 1 & 3 \end{pmatrix}, \text{ and } Q = 2 \begin{pmatrix} 1 & 0 & 1 \\ 0 & 0 & 0 \\ 1 & 0 & 1 \end{pmatrix},$$

where  $Q$  and  $P$  are respectively the weighting matrix and the solution matrix of the Lyapunov equation.. Check and state about the quality of the stability of  $A$  (asymptotic stable, boundary stable (stable), unstable).

**Solution:** The eigenvalues of  $P$  are  $2 \times (0.5858, 2.0000, 3.4142)$  and greater than zero, so  $P$  is a positive definite matrix.

The eigenvalues of  $Q$  are  $(0, 0, 2)$ , therefore  $Q$  is a semi-positive definite matrix.

Because the first derivative of the Lyapunov function  $V(t) = x^T P x$  is  $\dot{V}(t) = -x^T Q x$  and  $Q$  is a semi-positive definite matrix,  $\dot{V}(t) = -x^T Q x \leq 0$ . According to Lyapunov stability theory, it can be concluded that the system is stable.

From the given  $P$  and  $Q$  we can't get information about asymptotic stability. The eigenvectors of  $A$  are  $[1 \ -1 \ 1]^T$ ,  $[0.17 - 0.34i \ 0.85 - 0.17 + 0.34i]^T$ ,  $[0.17 + 0.34i \ 0.85 - 0.17 - 0.34i]^T$  and linear independent, therefore  $A$  can be transformed into a diagonal matrix. The eigenvalues of  $A$  is  $(-1, \pm i)$ , two of which lie in the imaginary axes. Therefore the system is boundary stable.

- e) Is the system described with  $A = \begin{bmatrix} 0 & 1 & 0 \\ a & 1 & a \\ 3 & 0 & 0 \end{bmatrix}$  stable? Calculate the stability condition(s) with respect to the parameter  $a$ ?

**Solution:** From  $\det(\lambda I - A) = 0$  the characteristic polynomial

$$\lambda^3 - \lambda^2 - a\lambda - 3a = 0$$

can be calculated.

The necessary condition for Hurwitz is not fulfilled ( $a_3 = 1, a_2 = -1$ ), so the system must be unstable.

• **Problem 2**

a) Calculate the damping of the eigenvalue pair  $\lambda_{1,2} = -1 \pm j$ .

**Solution:** The damping ratio is 0.707, calculated by  $\zeta = \cos(\arctan(\frac{1}{1}))$ .

b) The system

$$\dot{x} = Ax + Bu, y = Cx$$

with

$$A = \begin{pmatrix} 0 & 1 \\ -a & a-10 \end{pmatrix}, \quad B = b \begin{pmatrix} 1 \\ 2 \end{pmatrix}, \quad \text{and } C = c \begin{pmatrix} 10 & 1 \end{pmatrix}$$

is given. For which parameter  $a$  is the uncontrolled system asymptotic stable?

**Solution:** Condition for the state stability:  $Re\{\lambda_i\} \leq 0$ .

$$\det(\lambda I - A) = 0$$

$\implies$

$$\lambda^2 - (a-10)\lambda + a = 0$$

$\implies$

$$\lambda_{1,2} = \frac{(a-10) \pm \sqrt{(a-10)^2 - 4a}}{2}$$

If  $a \geq 0$ ,  $|a-10| \geq \sqrt{(a-10)^2 - 4a}$ ,  $Re\{\lambda_i\} \leq 0$  only requires  $a-10 < 0$ .

If  $a < 0$ ,  $|a-10| < \sqrt{(a-10)^2 - 4a}$  and therefore  $(a-10) + \sqrt{(a-10)^2 - 4a} > 0$ , which will lead to an unstable mode. So  $a$  must be greater than 0.

Therefore, when  $0 < a < 10$ , the uncontrolled system is stable.

c) What are the domains of parameters  $a, b$  if the system described in task b) is fully controllable?

**Solution:**  $Q_s = [B \ AB] = b \begin{bmatrix} 1 & 2 \\ 2 & a-20 \end{bmatrix}$ . Being fully controllable means  $rank(Q_s) = 2$ .

Therefore  $a-20 \neq 4$  and  $b \neq 0$ , which means  $a \neq 24$  and  $b \neq 0$ .

d) What are the domains of parameters  $a, c$  if the system described in task b) is fully observable?

**Solution:**  $Q_b = [C \ CA]^T = c \begin{bmatrix} 10 & 1 \\ -a & a \end{bmatrix}$ . Fully observable means  $rank(Q_b) = 2$ .

Therefore  $a \neq 0$  and  $c \neq 0$  must be fulfilled.

e) For which values of parameters  $k_1, k_2$  of a state feedback  $u = -k_1x_1 - k_2x_2$  has the controlled system described in task b) with  $a = 12, b = 1$  the eigenvalues  $\lambda_{1,2} = -2$

**Solution:**  $[k_1 \ k_2] = [\frac{14}{3} \ \frac{2}{3}]$ , calculated by pole placement.

# Problem 31

$$a) \det(\lambda I - A) = 0$$

$$\Leftrightarrow \lambda^3 + (-q-p)\lambda^2 + (-q-1+pq)\lambda + pq+p = 0$$

$$H = \begin{bmatrix} -q-p & (q+1)p & 0 \\ 1 & -q-1+pq & 0 \\ 0 & -q-p & (q+1)p \end{bmatrix}$$

HURWITZ

$$(1) \quad -q-p > 0 \Leftrightarrow -p > q \quad (1.1)$$

$$-q-1+pq > 0 \Leftrightarrow q(-1+p) > 1 \quad (1.2)$$

$$(q+1)p > 0 \quad (1.3)$$

$$(1.3): (q+1)p > 0 \Leftrightarrow \begin{cases} p > 0 \wedge q > -1 & (1.3a) \\ p < 0 \wedge q < -1 & (1.3b) \end{cases}$$

$$(1.2): a) q=0 \Rightarrow -1 > 0 \quad \text{⊗} \Rightarrow q \neq 0$$

$$b) p < 0 \Rightarrow q < \frac{1}{-1+p} < -1 \quad (\text{cf. (1.3b)})$$

$$c) 0 < p < 1 \Rightarrow -q < \frac{1}{-1+p} < -1 \quad \text{but } q > -1 \quad (1.3a) \quad \text{⊗}$$

$$d) p > 1 \Rightarrow -1+p > 0 \Rightarrow q > 0 \quad \text{but } 0 > -p > q \quad (1.1) \quad \text{⊗}$$

Problem 31

$$(1.1), (1.2), (1.3) \Rightarrow p < 0 \text{ and } q < -1 \text{ (necessary condition)}$$

$$(2) \det H_1 = -q - p > 0 \quad \checkmark \quad (\text{cf. (1.1)})$$

$$\det H_2 = q^2 + q - p q^2 - p^2 q > 0$$

$$\det H_3 = (q+1)p - \underbrace{\det H_2}_{> 0} > 0 \quad \checkmark \quad (\text{cf. (1.3)})$$

Calculate  $\det H_2$  for  $q < -1, p < 0$

$$q+1 - p q - p^2 < 0$$

$$\Leftrightarrow q(1-p) + 1 - p^2 < 0$$

$$\Leftrightarrow \underbrace{(1-p)}_{> 0} \underbrace{(q+p+1)}_{< 0} < 0$$

$$q+p+1 < 0 \Leftrightarrow q < -1-p < -1$$

This holds true for  $q < -1$

$\Rightarrow \{q < -1, p < 0\}$  is the necessary and sufficient condition for a state stable system.

### Problem 3)

$$b) A = \begin{bmatrix} 0 & -1 & 0 \\ -2 & -1 & 2 \\ 2 & -1 & -2 \end{bmatrix}$$

$$\det(\lambda I - A) = \lambda^3 + 3\lambda^2 + 2\lambda = 0 \Rightarrow \lambda_1 = 0, \lambda_2 = -1, \lambda_3 = -2$$

HAUPTUS for  $\lambda_1 = 0$ :

$$\text{Rang} \begin{bmatrix} 0 & 1 & 0 & 1 & 1+b \\ 2 & 1 & -2 & 1 & 1 \\ -2 & 1 & 2 & 1 & 1+b \end{bmatrix} = \text{Rang} \begin{bmatrix} 0 & 0 & 0 & 0 & b \\ 0 & 0 & 0 & 1 & 0 \\ 0 & 0 & 1 & 0 & 0 \end{bmatrix}$$

Full rank for  $b \neq 0 \Rightarrow$  For  $b \neq 0$  the system can be stabilized by state feedback.

$$c) \det(\lambda I - (A - BK)) \stackrel{!}{=} (\lambda + 1)(\lambda + 2)(\lambda + 16)$$

$$\Leftrightarrow \lambda^3 + (19 + k_{12} + k_{22})\lambda^2 + (50 + 18k_{12} + 18k_{22})\lambda + 32k_{12} + 32k_{22} + 32$$

$$\stackrel{!}{=} \lambda^3 + 19\lambda^2 + 50\lambda + 32$$

$$\Leftrightarrow k_{12} = -k_{22}$$

d) HAUPTUS for  $\lambda_3 = -2$ :

$$\text{Rang} \begin{bmatrix} -2 & 1 & 0 & 1 & 2 \\ 2 & -1 & -2 & 1 & 1 \\ -2 & 1 & 0 & 1 & 2 \end{bmatrix} = \text{Rang} \begin{bmatrix} 0 & 0 & 0 & 1 & 0 \\ 0 & 0 & 1 & 0 & 0 \\ 0 & 0 & 0 & 1 & 0 \end{bmatrix} = 2$$

$\Rightarrow \lambda_3 = -2$  is not controllable  $\Rightarrow$  the eigenvalues  $\{0, -1, -2\}$  cannot be shifted to  $\{0, -1, -4\}$ .

Problem 21  
e) Hauptvektoren für  $\lambda_2 = -1$ :

$$\text{Rank} \begin{bmatrix} -1 & 1 & 0 & 1 & 2 \\ 2 & 0 & -2 & 1 & 1 \\ -2 & 1 & 1 & 1 & 2 \end{bmatrix} = \text{Rank} \begin{bmatrix} 0 & 1 & 0 & 0 & 0 \\ 0 & 0 & 0 & 0 & 1 \\ 0 & 0 & 1 & 0 & 0 \end{bmatrix} = 3$$

$\Rightarrow \lambda_2 = -1$  is controllable

The system is fully observable (assumption given in the problem description) and the eigenvalues  $\lambda_1 = 0$  and  $\lambda_2 = -1$  are controllable. So the poles are  $\{0, -1\}$ .

$\lambda_3 = -2$  is no pole.

$$a) G(s) = C (sI - A)^{-1} B + D$$

$$(sI - A)^{-1} = \frac{1}{\det(sI - A)} \begin{bmatrix} s+20 & -8 \\ -8 & s+8 \end{bmatrix}$$

$$= \frac{1}{(s+8)(s+20)-64} \begin{bmatrix} s+20 & -8 \\ -8 & s+8 \end{bmatrix}$$

$$G(s) = \begin{bmatrix} 0 & 1 \\ 1 & 0 \end{bmatrix} \frac{1}{(s+8)(s+20)-64} \begin{bmatrix} s+20 & -8 \\ -8 & s+8 \end{bmatrix} \begin{bmatrix} 1 \\ -1 \end{bmatrix}$$

$$= \frac{1}{(s+8)(s+20)-64} \begin{bmatrix} 0 & 1 \\ 1 & 0 \end{bmatrix} \begin{bmatrix} s+20+8 \\ -8-s-8 \end{bmatrix} = \frac{1}{(s+8)(s+20)-64} \begin{bmatrix} -s-16 \\ s+28 \end{bmatrix}$$

$$\text{poles: } s^2 + 28s + 96 = 0$$

$$s_{1,2} = -14 \pm \sqrt{196 - 96} = -14 \pm 10$$

$$s_1 = -24 \quad s_2 = -4$$

$$\text{zeros: } -s_{01} - 16 = 0$$

$$s_{01} = -16$$

$$s_{02} + 28 = 0$$

$$s_{02} = -28$$

$$b) Q_s = [B, AB] = \begin{bmatrix} 1 & 0 \\ -1 & 12 \end{bmatrix}; \quad \text{rank } Q_s = 2 \Rightarrow \text{fully contr.}$$

$$c) Q_B = \begin{bmatrix} C \\ CA \end{bmatrix} = \begin{bmatrix} 0 & 1 \\ 1 & 0 \\ -8 & -20 \\ -8 & -8 \end{bmatrix}; \quad \text{rank } (Q_B) = 2 \Rightarrow \text{fully observable}$$



$$\begin{aligned}
 d) \quad |\lambda I - (A - BK)| &= \left| \lambda I - \left( \begin{bmatrix} -8 & -8 \\ -8 & -20 \end{bmatrix} - \begin{bmatrix} 1 \\ -1 \end{bmatrix} \begin{bmatrix} k_1 & k_2 \end{bmatrix} \right) \right| \\
 &= \left| \lambda I - \left( \begin{bmatrix} -8 & -8 \\ -8 & -20 \end{bmatrix} - \begin{bmatrix} k_1 & k_2 \\ -k_1 & -k_2 \end{bmatrix} \right) \right| \\
 &= \left| \lambda I - \begin{bmatrix} -8-k_1 & -8-k_2 \\ -8+k_1 & -20+k_2 \end{bmatrix} \right| = \begin{vmatrix} \lambda+8+k_1 & 8+k_2 \\ 8-k_1 & \lambda+20-k_2 \end{vmatrix} \\
 &= \lambda^2 + \lambda(k_1 - k_2 + 28) + 28k_1 - 16k_2 - 96
 \end{aligned}$$

$$k_1 - k_2 + 28 > 0 \quad (I)$$

$$28k_1 - 16k_2 - 96 > 0 \quad (II)$$

$$\text{ans I: } k_1 + 28 > k_2 \quad \Leftrightarrow \quad k_1 > k_2 - 28$$

$$\text{ans II: } 28k_1 - 96 > 16k_2$$

$$\frac{28}{16}k_1 - \frac{96}{16} > k_2$$

$$k_1 > \left( k_2 + \frac{96}{16} \right) \frac{16}{28}$$

$$k_1 > \left( k_2 \cdot \frac{16}{28} + \frac{96}{28} \right)$$

$$k_1 > k_2 \cdot \frac{4}{7} + \frac{24}{7}$$

$$Q = \begin{bmatrix} 84 & 96 \\ 96 & 112 \end{bmatrix} \quad R = 0,25$$

$$A = \begin{bmatrix} -8 & -8 \\ -8 & -20 \end{bmatrix}; \quad B = \begin{bmatrix} 1 \\ -1 \end{bmatrix}; \quad P = \begin{bmatrix} p_1 & p_2 \\ p_2 & p_2 \end{bmatrix}$$

$$A^T P + P A - P B R^{-1} B^T P = -Q$$

$$\begin{bmatrix} -8 & -8 \\ -8 & -20 \end{bmatrix} \begin{bmatrix} p_1 & p_2 \\ p_2 & p_2 \end{bmatrix} + \begin{bmatrix} p_1 & p_2 \\ p_2 & p_2 \end{bmatrix} \begin{bmatrix} -8 & -8 \\ -8 & -20 \end{bmatrix} \dots$$

$$- \begin{bmatrix} p_1 & p_2 \\ p_2 & p_2 \end{bmatrix} \begin{bmatrix} 1 \\ -1 \end{bmatrix} \cdot 4 \cdot \begin{bmatrix} 1 \\ -1 \end{bmatrix} \begin{bmatrix} p_1 & p_2 \\ p_2 & p_2 \end{bmatrix}$$

$$= \begin{bmatrix} -8p_1 - 8p_2 & -8p_2 - 8p_2 \\ -8p_1 - 20p_2 & -8p_2 - 20p_2 \end{bmatrix} + \begin{bmatrix} -8p_1 - 8p_2 & -8p_1 - 20p_2 \\ -8p_2 - 8p_2 & -8p_2 - 20p_2 \end{bmatrix} - \left( \begin{bmatrix} p_1 - p_2 \\ p_2 - p_2 \end{bmatrix} \cdot 4 \cdot \begin{bmatrix} p_1 - p_2 \\ p_2 - p_2 \end{bmatrix} \right)$$

$$= \begin{bmatrix} -8p_1 - 8p_2 & -16p_2 \\ -8p_1 - 20p_2 & -28p_2 \end{bmatrix} + \begin{bmatrix} -8p_1 - 8p_2 & -8p_1 - 20p_2 \\ -16p_2 & -28p_2 \end{bmatrix} - 4 \begin{bmatrix} (p_1 - p_2)^2 & 0 \\ 0 & 0 \end{bmatrix}$$

$$= \begin{bmatrix} 16p_1 + 16p_2 - 4(p_1 - p_2)^2 & 8p_1 + 36p_2 \\ 8p_1 + 36p_2 & 56p_2 \end{bmatrix} = \begin{bmatrix} 84 & 96 \\ 96 & 112 \end{bmatrix}$$

$$56p_2 = 112$$

$$p_2 = 2$$

$$8p_1 + 72 = 96$$

$$8p_1 = 24$$

$$p_1 = 3$$

$$K^* = R^{-1} B^T \cdot P = 4 \begin{bmatrix} 1 & -1 \end{bmatrix} \begin{bmatrix} 3 & 2 \\ 2 & 2 \end{bmatrix}$$

$$= \begin{bmatrix} 4 & -4 \end{bmatrix} \begin{bmatrix} 3 & 2 \\ 2 & 2 \end{bmatrix} = \begin{bmatrix} 12 & -8 & 0 \end{bmatrix} = \begin{bmatrix} 4 & 0 \end{bmatrix}$$

$$f) \text{rank} \begin{bmatrix} c \\ c(A-BK) \end{bmatrix} = \text{rank} \begin{bmatrix} 0 & 1 \\ 1 & 0 \\ c \left( \begin{bmatrix} -8 & -8 \\ -8 & -20 \end{bmatrix} - \begin{bmatrix} 1 \\ -1 \end{bmatrix} \begin{bmatrix} k_1 & k_2 \end{bmatrix} \right) \end{bmatrix}$$

$$= \text{rank} \begin{bmatrix} 0 & 1 \\ 1 & 0 \\ \begin{bmatrix} 0 & 1 \\ 1 & 0 \end{bmatrix} \cdot \begin{bmatrix} -8-k_1 & -8-k_2 \\ -8+k_1 & -20+k_2 \end{bmatrix} \end{bmatrix}$$

$$= \text{rank} \begin{bmatrix} 0 & 1 \\ 1 & 0 \\ -8+k_1 & -20+k_2 \\ -8-k_1 & -8-k_2 \end{bmatrix} = 2 \Rightarrow \text{always observable}$$